

# DSP 2 HELPSHEETS

## Signals:

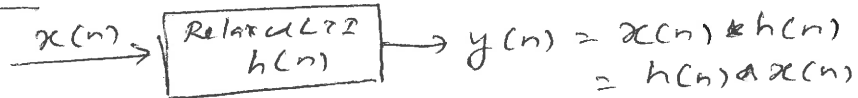
unit pulse/impulse:

$$\delta(n) = \begin{cases} 1, & n=0 \\ 0, & n \neq 0 \end{cases}$$

unit step:

$$u(n) = \begin{cases} 1, & n \geq 0 \\ 0, & n < 0 \end{cases}$$

## LTI System:



$$y(n] = \sum_{k=-\infty}^{\infty} x(k] h(n-k] = \sum_{k=-\infty}^{\infty} h(k] x(n-k], \quad -\infty < n < \infty$$

## DT convolution Properties:

- $x(n] * \delta(n-n_0] = x(n-n_0]$
- $x(n-d] * h(n-\beta] = y(n-d-\beta]$
- Step Response:  $y_s(n] = \sum_{k=-\infty}^n h(k]$

## Geometric Series Formulas:

$$\sum_{n=0}^{N-1} d^n = \frac{1-d^N}{1-d}, \quad \sum_{n=0}^{\infty} d^n = \frac{1}{1-d}, \quad |d| < 1$$

## Correlation:

$$r_{xy}(l] = \sum_{n=-\infty}^{\infty} x(n] y(n-l] = x(l] * y(-l]$$

## Periodic Correlation:

$$r_{xx}(l] = \frac{1}{N} \sum_{n=0}^{N-1} x(n] x(n-l] = \frac{1}{N} \cdot x(l] \circledast x(-l]$$

$$r_{xy}(l] = \frac{1}{N} \sum_{n=0}^{N-1} x(n] y(n-l] = \frac{1}{N} \cdot x(l] \circledast y(-l]$$

## Circular/Periodic Convolution

$$x(n] \circledast h(n] = \sum_{k=0}^{N-1} x(k] h((n-k))_N = \sum_{k=0}^{N-1} h(k] x((n-k))_N$$

# Z-TRANSFORM TABLE & PROPERTIES

TABLE 3.2 Properties of the z-Transform

Property	Time Domain	z-Domain	ROC
Notation	$x(n)$	$X(z)$	ROC: $r_2 <  z  < r_1$
	$x_1(n)$	$X_1(z)$	ROC <sub>1</sub>
	$x_2(n)$	$X_2(z)$	ROC <sub>2</sub>
Linearity	$a_1x_1(n) + a_2x_2(n)$	$a_1X_1(z) + a_2X_2(z)$	At least the intersection of ROC <sub>1</sub> and ROC <sub>2</sub>
Time shifting	$x(n-k)$	$z^{-k}X(z)$	That of $X(z)$ , except $z=0$ if $k > 0$ and $z=\infty$ if $k < 0$
Scaling in the z-domain	$a^n x(n)$	$X(a^{-1}z)$	$ a r_2 <  z  <  a r_1$
Time reversal	$x(-n)$	$X(z^{-1})$	$\frac{1}{r_1} <  z  < \frac{1}{r_2}$
Conjugation	$x^*(n)$	$X^*(z^*)$	ROC
Real part	$\text{Re}\{x(n)\}$	$\frac{1}{2}[X(z) + X^*(z^*)]$	Includes ROC
Imaginary part	$\text{Im}\{x(n)\}$	$\frac{1}{2j}[X(z) - X^*(z^*)]$	Includes ROC
Differentiation in the z-domain	$nx(n)$	$-z \frac{dX(z)}{dz}$	$r_2 <  z  < r_1$
Convolution	$x_1(n) * x_2(n)$	$X_1(z)X_2(z)$	At least, the intersection of ROC <sub>1</sub> and ROC <sub>2</sub>
Correlation	$r_{x_1x_2}(l) = x_1(l) * x_2(-l)$	$R_{x_1x_2}(z) = X_1(z)X_2(z^{-1})$	At least, the intersection of ROC of $X_1(z)$ and $X_2(z^{-1})$
Initial value theorem	If $x(n)$ causal	$x(0) = \lim_{z \rightarrow \infty} X(z)$	
Multiplication	$x_1(n)x_2(n)$	$\frac{1}{2\pi j} \oint_C X_1(v)X_2\left(\frac{z}{v}\right)v^{-1}dv$	At least, $r_{11}r_{21} <  z  < r_{1u}r_{2u}$
Parseval's relation	$\sum_{n=-\infty}^{\infty} x_1(n)x_2^*(n)$	$\frac{1}{2\pi j} \oint_C X_1(v)X_2^*(1/v^*)v^{-1}dv$	

TABLE 3.3 Some Common z-Transform Pairs

Signal, $x(n)$	z-Transform, $X(z)$	ROC
1 $\delta(n)$	1	All $z$
2 $u(n)$	$\frac{1}{1-z^{-1}}$	$ z  > 1$
3 $a^n u(n)$	$\frac{1}{1-az^{-1}}$	$ z  >  a $
4 $na^n u(n)$	$\frac{az^{-1}}{(1-az^{-1})^2}$	$ z  >  a $
5 $-a^n u(-n-1)$	$\frac{1}{1-az^{-1}}$	$ z  <  a $
6 $-na^n u(-n-1)$	$\frac{az^{-1}}{(1-az^{-1})^2}$	$ z  <  a $
7 $(\cos \omega_0 n)u(n)$	$\frac{1-z^{-1} \cos \omega_0}{1-2z^{-1} \cos \omega_0 + z^{-2}}$	$ z  > 1$
8 $(\sin \omega_0 n)u(n)$	$\frac{z^{-1} \sin \omega_0}{1-2z^{-1} \cos \omega_0 + z^{-2}}$	$ z  > 1$
9 $(a^n \cos \omega_0 n)u(n)$	$\frac{1-az^{-1} \cos \omega_0}{1-2az^{-1} \cos \omega_0 + a^2 z^{-2}}$	$ z  >  a $
10 $(a^n \sin \omega_0 n)u(n)$	$\frac{az^{-1} \sin \omega_0}{1-2az^{-1} \cos \omega_0 + a^2 z^{-2}}$	$ z  >  a $

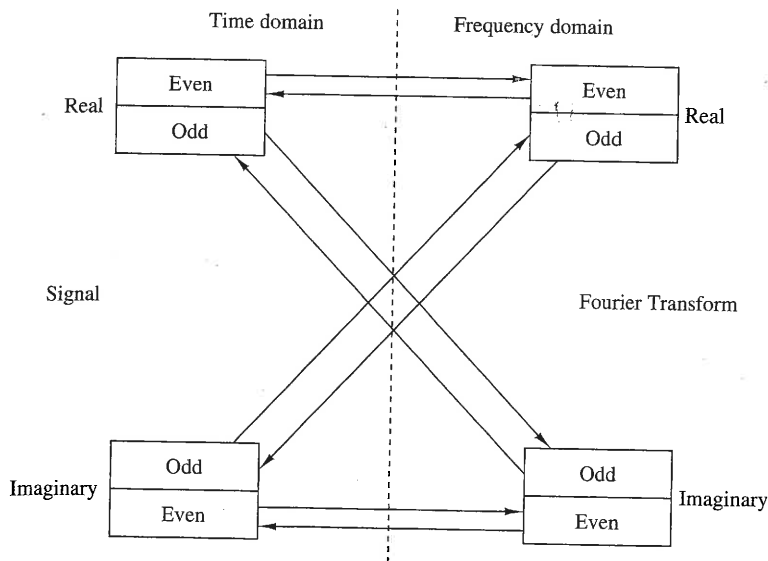
# DTFT SYMMETRY PROPERTIES

**TABLE 4.4** Symmetry Properties of the Discrete-Time Fourier Transform

Sequence	DTFT
$x(n)$	$X(\omega)$
$x^*(n)$	$X^*(-\omega)$
$x^*(-n)$	$X^*(\omega)$
$x_R(n)$	$X_e(\omega) = \frac{1}{2}[X(\omega) + X^*(-\omega)]$
$jx_I(n)$	$X_o(\omega) = \frac{1}{2}[X(\omega) - X^*(-\omega)]$
$x_e(n) = \frac{1}{2}[x(n) + x^*(-n)]$	$X_R(\omega)$
$x_o(n) = \frac{1}{2}[x(n) - x^*(-n)]$	$jX_I(\omega)$

<p style="text-align: center;">Real Signals</p> <p>Any real signal <math>x(n)</math></p> <p><math>x_e(n) = \frac{1}{2}[x(n) + x(-n)]</math> (real and even)</p> <p><math>x_o(n) = \frac{1}{2}[x(n) - x(-n)]</math> (real and odd)</p>	<p style="text-align: center;">Real Signals</p> <p><math>X(\omega) = X^*(-\omega)</math></p> <p><math>X_R(\omega) = X_R(-\omega)</math></p> <p><math>X_I(\omega) = -X_I(-\omega)</math></p> <p><math> X(\omega)  =  X(-\omega) </math></p> <p><math>\angle X(\omega) = -\angle X(-\omega)</math></p> <p><math>X_R(\omega)</math> (real and even)</p> <p><math>jX_I(\omega)</math> (imaginary and odd)</p>
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**Figure 4.4.2** Summary of symmetry properties for the Fourier transform.

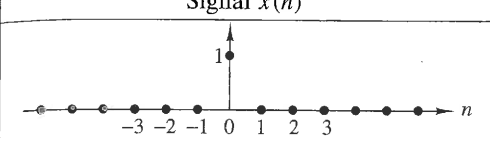
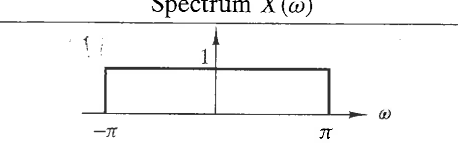
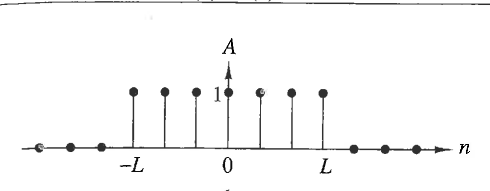
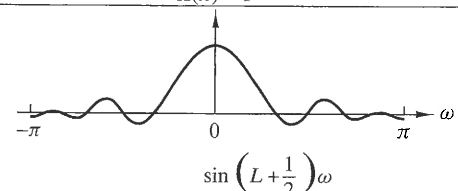
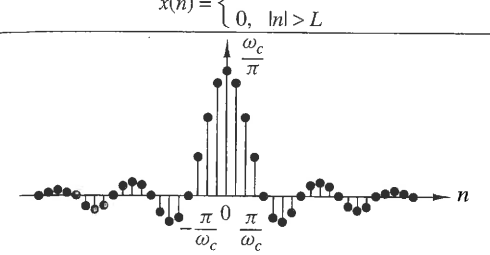
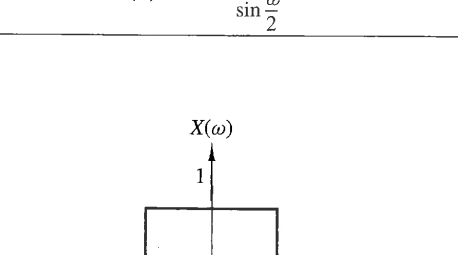
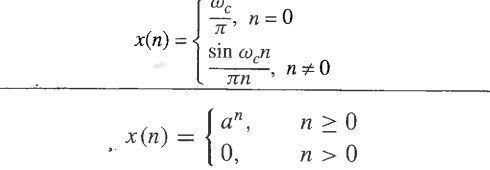
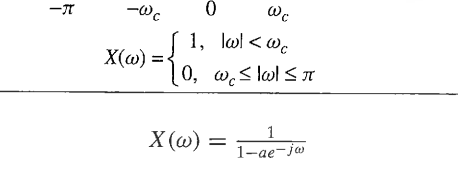
# DTFT: Table & Properties

$$X(\omega) = \sum_{n=-\infty}^{\infty} x(n) e^{-j\omega n}$$

**TABLE 4.5** Properties of the Fourier Transform for Discrete-Time Signals

Property	Time Domain	Frequency Domain
Notation	$x(n)$	$X(\omega)$
	$x_1(n)$	$X_1(\omega)$
	$x_2(n)$	$X_2(\omega)$
Linearity	$a_1x_1(n) + a_2x_2(n)$	$a_1X_1(\omega) + a_2X_2(\omega)$
Time shifting	$x(n - k)$	$e^{-j\omega k} X(\omega)$
Time reversal	$x(-n)$	$X(-\omega)$
Convolution	$x_1(n) * x_2(n)$	$X_1(\omega)X_2(\omega)$
Correlation	$r_{x_1x_2}(l) = x_1(l) * x_2(-l)$	$S_{x_1x_2}(\omega) = X_1(\omega)X_2(-\omega)$
		$= X_1(\omega)X_2^*(\omega)$ [if $x_2(n)$ is real]
Wiener-Khinchine theorem	$r_{xx}(l)$	$S_{xx}(\omega)$
Frequency shifting	$e^{j\omega_0 n} x(n)$	$X(\omega - \omega_0)$
Modulation	$x(n) \cos \omega_0 n$	$\frac{1}{2} X(\omega + \omega_0) + \frac{1}{2} X(\omega - \omega_0)$
Multiplication	$x_1(n)x_2(n)$	$\frac{1}{2\pi} \int_{-\pi}^{\pi} X_1(\lambda)X_2(\omega - \lambda) d\lambda$
Differentiation in the frequency domain	$nx(n)$	$j \frac{dX(\omega)}{d\omega}$
		$X^*(-\omega)$
Conjugation	$x^*(n)$	$X^*(-\omega)$
Parseval's theorem	$\sum_{n=-\infty}^{\infty} x_1(n)x_2^*(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X_1(\omega)X_2^*(\omega) d\omega$	

**TABLE 4.6** Some Useful Fourier Transform Pairs for Discrete-Time Aperiodic Signals

Signal $x(n)$	Spectrum $X(\omega)$
 <p><math>x(n) = \delta(n)</math></p>	 <p><math>X(\pi) = 1</math></p>
 <p><math>x(n) = \begin{cases} A, &amp;  n  \leq L \\ 0, &amp;  n  &gt; L \end{cases}</math></p>	 <p><math>X(\omega) = A \frac{\sin\left(L + \frac{1}{2}\right)\omega}{\sin\frac{\omega}{2}}</math></p>
 <p><math>x(n) = \begin{cases} \frac{\omega_c}{\pi}, &amp; n = 0 \\ \frac{\sin \omega_c n}{\pi n}, &amp; n \neq 0 \end{cases}</math></p>	 <p><math>X(\omega) = \begin{cases} 1, &amp;  \omega  &lt; \omega_c \\ 0, &amp; \omega_c \leq  \omega  \leq \pi \end{cases}</math></p>
 <p><math>x(n) = \begin{cases} a^n, &amp; n \geq 0 \\ 0, &amp; n &lt; 0 \end{cases}</math></p>	 <p><math>X(\omega) = \frac{1}{1 - ae^{-j\omega}}</math></p>

# DFT

## DFT PAIR:

$$x(n) \longleftrightarrow X(k)$$

$$X(k) = \sum_{n=0}^{N-1} x(n) e^{-j\frac{2\pi}{N}kn}$$

$$x(n) = \frac{1}{N} \sum_{k=0}^{N-1} X(k) e^{j\frac{2\pi}{N}kn}$$

TABLE 7.1 Symmetry Properties of the DFT

$N$ -Point Sequence $x(n)$ , $0 \leq n \leq N-1$	$N$ -Point DFT
$x(n)$	$X(k)$
$x^*(n)$	$X^*(N-k)$
$x^*(N-n)$	$X^*(k)$
$x_R(n)$	$X_{ce}(k) = \frac{1}{2}[X(k) + X^*(N-k)]$
$jX_I(n)$	$X_{co}(k) = \frac{1}{2}[X(k) - X^*(N-k)]$
$x_{ce}(n) = \frac{1}{2}[x(n) + x^*(N-n)]$	$X_R(k)$
$x_{co}(n) = \frac{1}{2}[x(n) - x^*(N-n)]$	$jX_I(k)$
Real Signals	
Any real signal	$X(k) = X^*(N-k)$
$x(n)$	$X_R(k) = X_R(N-k)$
	$X_I(k) = -X_I(N-k)$
	$ X(k)  =  X(N-k) $
	$\angle X(k) = -\angle X(N-k)$
$x_{ce}(n) = \frac{1}{2}[x(n) + x(N-n)]$	$X_R(k)$
$x_{co}(n) = \frac{1}{2}[x(n) - x(N-n)]$	$jX_I(k)$

TABLE 7.2 Properties of the DFT

Property	Time Domain	Frequency Domain
Notation	$x(n), y(n)$	$X(k), Y(k)$
Periodicity	$x(n) = x(n+N)$	$X(k) = X(k+N)$
Linearity	$a_1x_1(n) + a_2x_2(n)$	$a_1X_1(k) + a_2X_2(k)$
Time reversal	$x(N-n)$	$X(N-k)$
Circular time shift	$x((n-l))_N$	$X(k)e^{-j2\pi kl/N}$
Circular frequency shift	$x(n)e^{j2\pi ln/N}$	$X((k-l))_N$
Complex conjugate	$x^*(n)$	$X^*(N-k)$
Circular convolution	$x_1(n) \otimes x_2(n)$	$X_1(k)X_2(k)$
Circular correlation	$x(n) \otimes y^*(-n)$	$X(k)Y^*(k)$
Multiplication of two sequences	$x_1(n)x_2(n)$	$\frac{1}{N}X_1(k) \otimes X_2(k)$
Parseval's theorem	$\sum_{n=0}^{N-1} x(n)y^*(n)$	$\frac{1}{N} \sum_{k=0}^{N-1} X(k)Y^*(k)$

## Upsampling & Downsampling

upsampling

$$* \quad y(n) = \begin{cases} x\left(\frac{n}{L}\right), & n = kL, \quad k = 0, \pm 1, \pm 2, \dots \\ 0, & \text{otherwise} \end{cases}$$

$$\Rightarrow \quad Y(\omega) = X(\omega L) \\ Y(z) = X(z^L)$$

Downsampling

$$* \quad y(n) = x(Mn)$$

$$Y(\omega) = \frac{1}{M} \sum_{k=0}^{M-1} X\left(\frac{\omega - 2\pi k}{M}\right)$$